

# ECEN689: Special Topics in High-Speed Links Circuits and Systems Spring 2010

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## Lecture 23: Jitter



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# Announcements

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- HW6 due Wednesday April 7 (in class)
- Exam 2 will be either April 28 or 30
- Reading
  - Will post some jitter application notes
  - Majority of today's material from Hall reference

# Agenda

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- Noise Budget Example
- Jitter

# Noise Source Classifications

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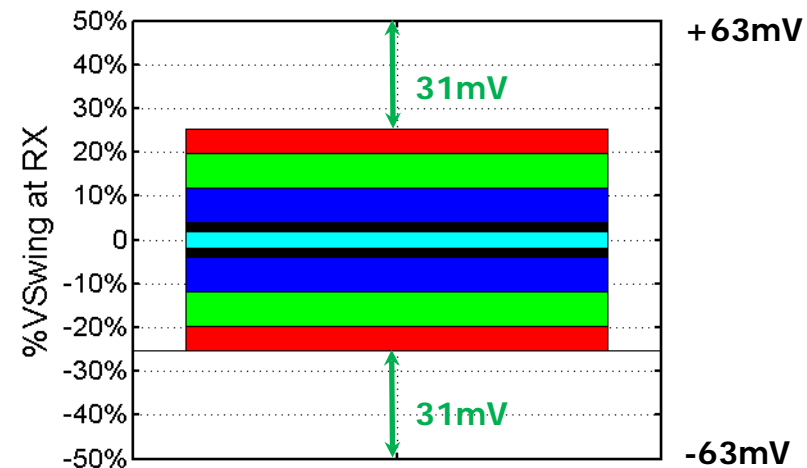
- Determining whether noise source is
  - Proportional vs Independent
  - Bounded vs Statistical
- is important in noise budgeting

	<b>Proportional</b>	<b>Independent</b>
<b>Bounded</b>	<b>Residual ISI</b> <b>Crosstalk</b>	<b>RX Offset</b> <b>RX Sensitivity</b> <b>Power Supply Noise</b>
<b>Statistical</b>	<b>Large-Channel Crosstalk</b>	<b>Random Noise</b>

# Noise Budget Example

- Peak TX differential swing of  $400\text{mV}_{\text{ppd}}$  equalized down 10dB
  - $\pm 200\text{mV} \rightarrow \pm 63\text{mV}$

Parameter	$K_n$	RMS	Value (BER = $10^{-12}$ )
Peak Differential Swing			0.4V
RX Offset + Sensitivity			5mV
Power Supply Noise			5mV
Residual ISI	0.05		20mV
Crosstalk	0.05		20mV
Random Noise		1mV	14mV
Attenuation	10dB = 0.684		0.274V
Total Noise			0.338V
Differential Eye Height Margin			62mV

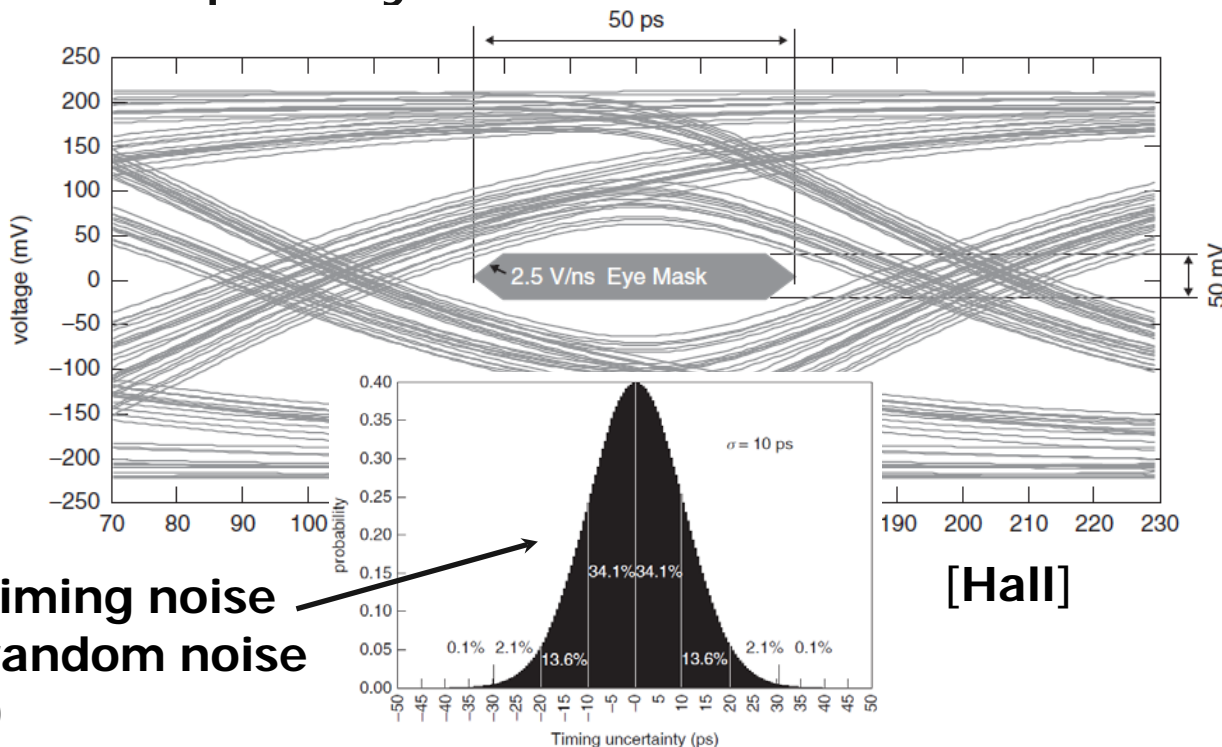


- Conservative analysis
  - Assumes all distributions combine at worst-case
- Better technique is to use statistical BER link simulators

# Eye Diagram and Spec Mask

- Links must have margin in both the voltage AND timing domain for proper operation
- For independent design (interoperability) of TX and RX, a spec eye mask is used

Eye at RX sampler



RX clock timing noise or jitter (random noise only here)

# Jitter Definitions

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- Jitter can be defined as “the short-term variation of a signal with respect to its ideal position in time”
- Jitter measurements
  - Period Jitter ( $J_{PER}$ )
    - Time difference between measured period and ideal period
  - Cycle to Cycle Jitter ( $J_{CC}$ )
    - Time difference between two adjacent clock periods
    - Important for budgeting on-chip digital circuits cycle time
  - Accumulated Jitter ( $J_{AC}$ )
    - Time difference between measured clock and ideal trigger clock
    - Jitter measurement most relative to high-speed link systems

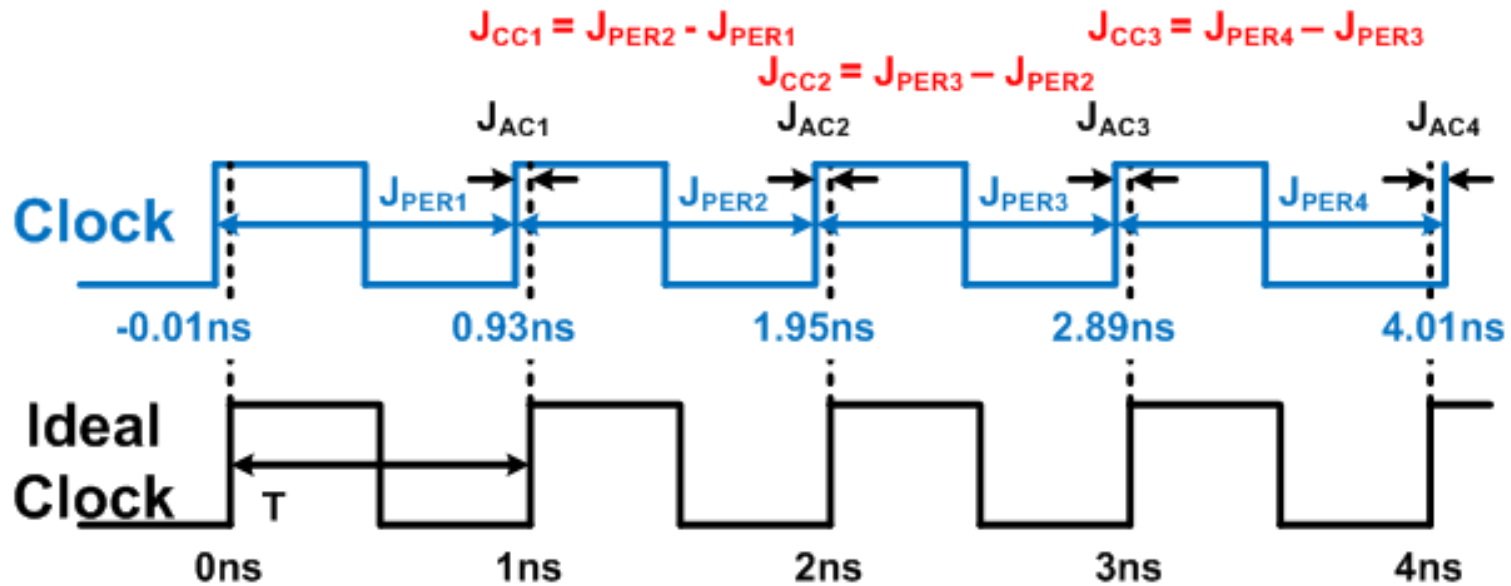
# Jitter Statistical Parameters

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- Mean Value
  - Can be interpreted as a fixed timing offset or “skew”
  - Generally not important, as usually can corrected for
- RMS Jitter
  - Useful for characterizing random component of jitter
- Peak-to-Peak Jitter
  - Function of both deterministic (bounded) and random (unbounded) jitter components
  - Must be quoted at a given BER to account for random (unbounded) jitter

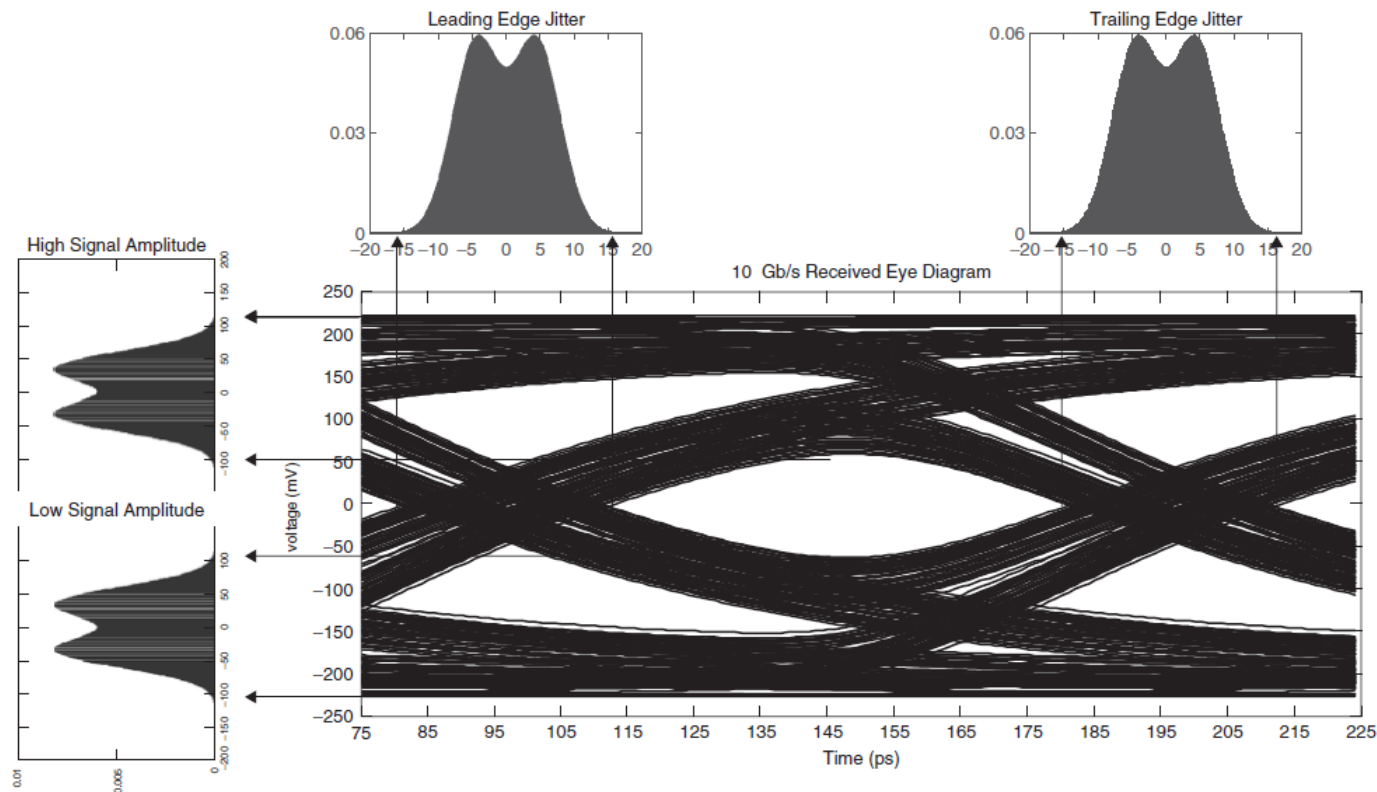


# Jitter Calculation Examples



n	1	2	3	4	Mean	RMS	PP
$J_{PER}$	-0.06	0.02	-0.06	0.12	0.005	0.085	0.18
$J_{CC}$	0.08	-0.08	0.18	-	0.06	0.131	0.26
$J_{AC}$	-0.07	-0.05	-0.11	0.01	-0.055	0.05	0.12

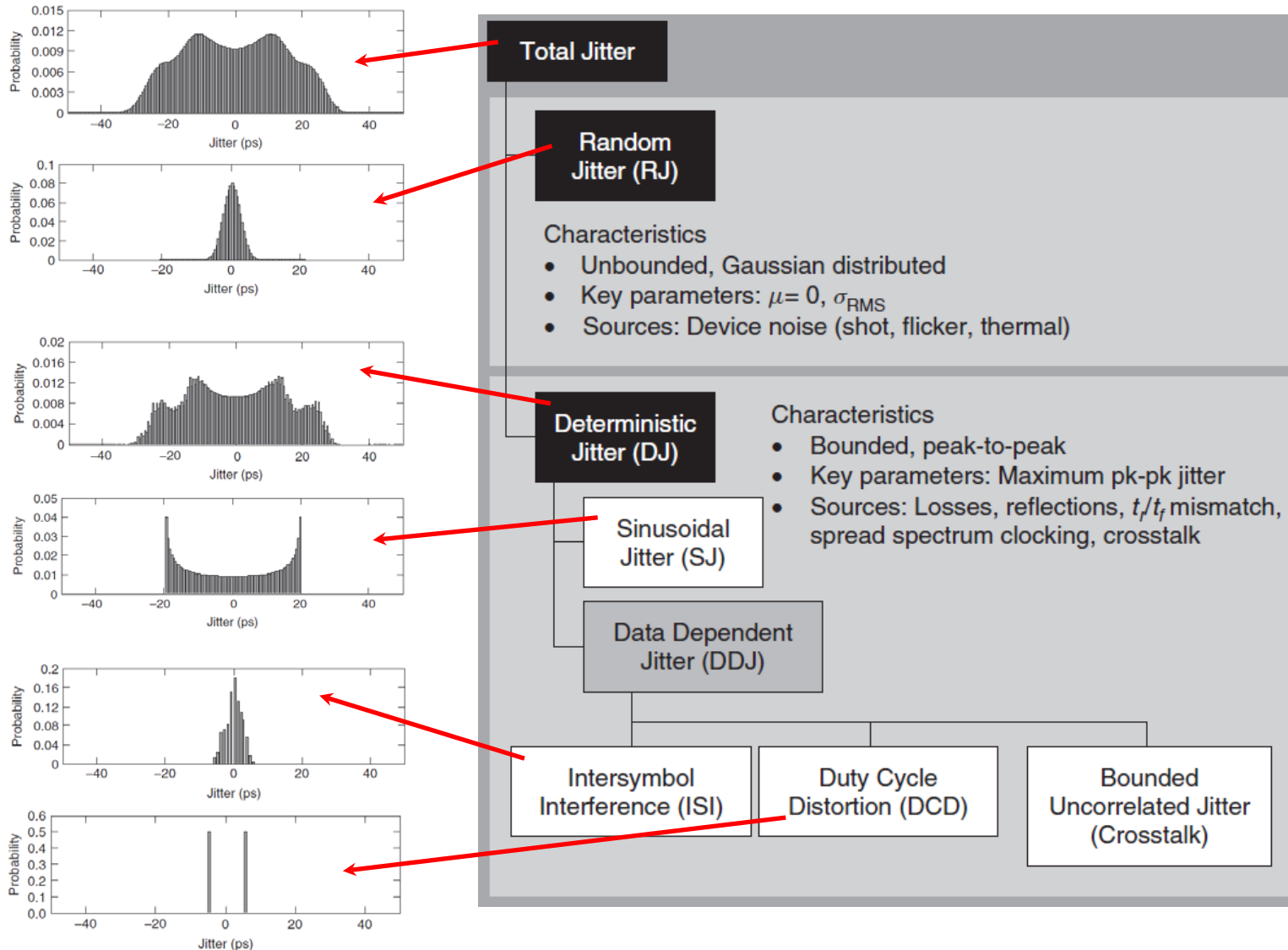
# Jitter Histogram



[Hall]

- Used to extract the jitter PDF
- Consists of both deterministic and random components
  - Need to decompose these components to accurately estimate jitter at a given BER

# Jitter Categories

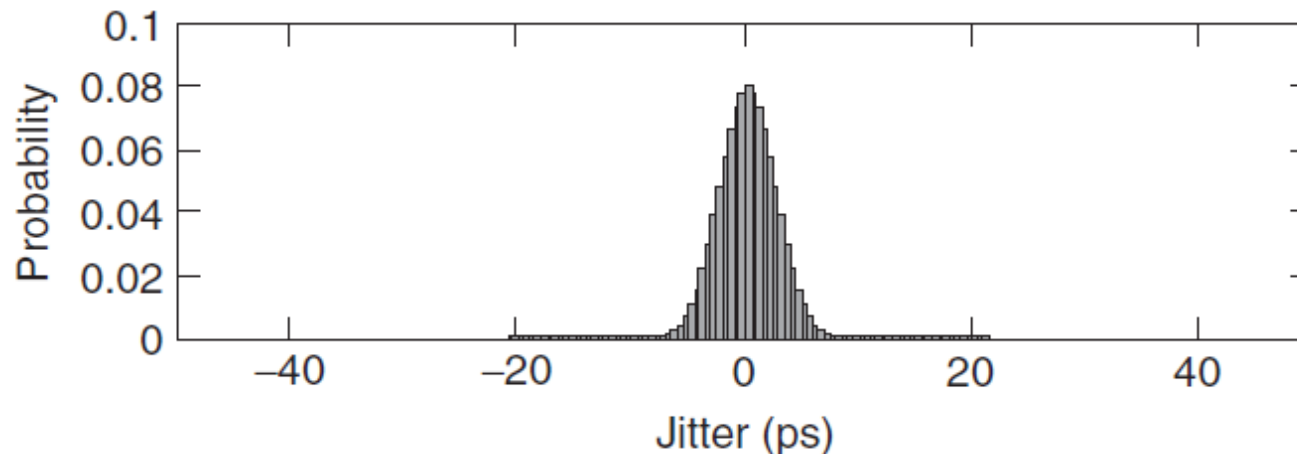


# Random Jitter (RJ)

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- Unbounded and modeled with a gaussian distribution
  - Assumed to have zero mean value
  - Characterized by the rms value,  $\sigma_{RJ}$
  - Peak-to-peak value must be quoted at a given BER
- Originates from device noise
  - Thermal, shot, flicker noise

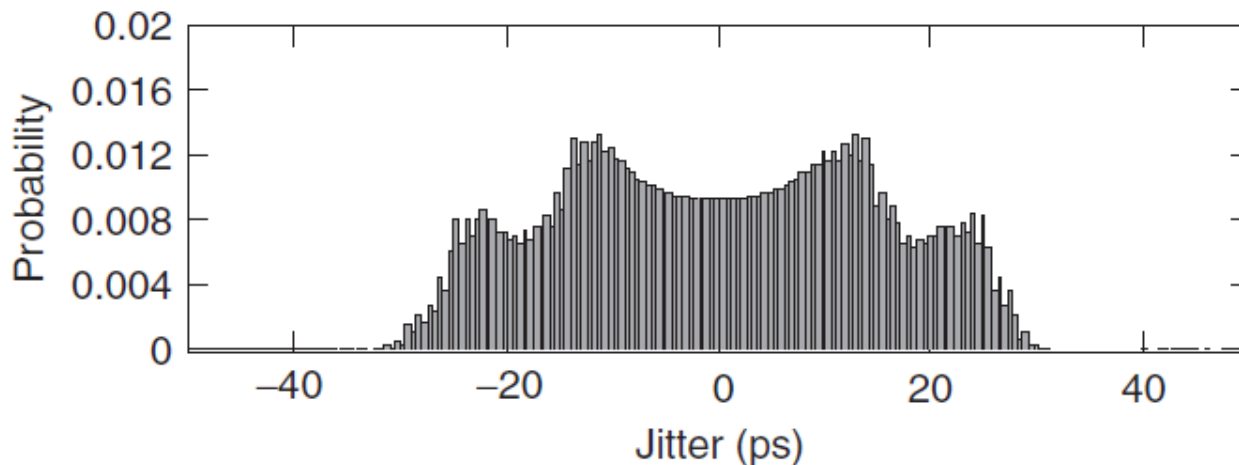
$$RJ(t) = \frac{1}{\sqrt{2\pi}\sigma_{RJ}} e^{\frac{-t^2}{2\sigma_{RJ}^2}}$$



# Deterministic Jitter (DJ)

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- Bounded with a peak-to-peak value that can be predicted
- Caused by transmission-line losses, duty-cycle distortion, spread-spectrum clocking, crosstalk
- Categories
  - Sinusoidal Jitter (SJ or PJ)
  - Data Dependent Jitter (DDJ)
    - Intersymbol Interference (ISI)
    - Duty Cycle Distortion (DCD)
    - Bounded Uncoirrelated Jitter (BUJ)



# Sinusoidal or Periodic Jitter (SJ or PJ)

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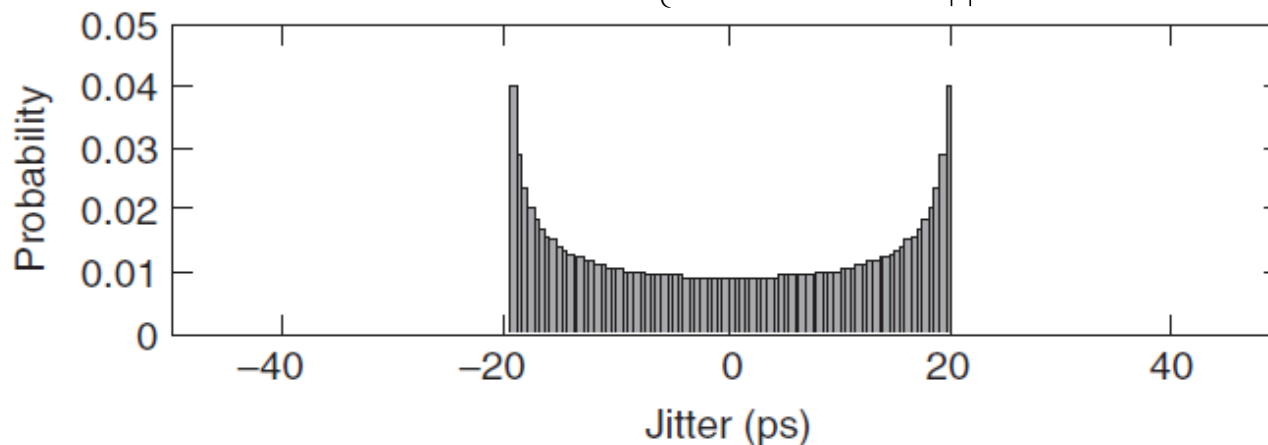
- Repeats at a fixed frequency due to modulating effects
  - Spread spectrum clocking
  - PLL reference clock feedthrough

- Can be decomposed into a Fourier series of sinusoids

$$SJ(t) = \sum_i A_i \cos(\omega_i t + \theta_i)$$

- The jitter produced by an individual sinusoid is

$$PDF_{SJ}(t) = \begin{cases} \frac{1}{\pi\sqrt{A^2 - t^2}} & A > |t| \\ 0 & A \leq |t| \end{cases}$$



# Data Dependent Jitter (DDJ)

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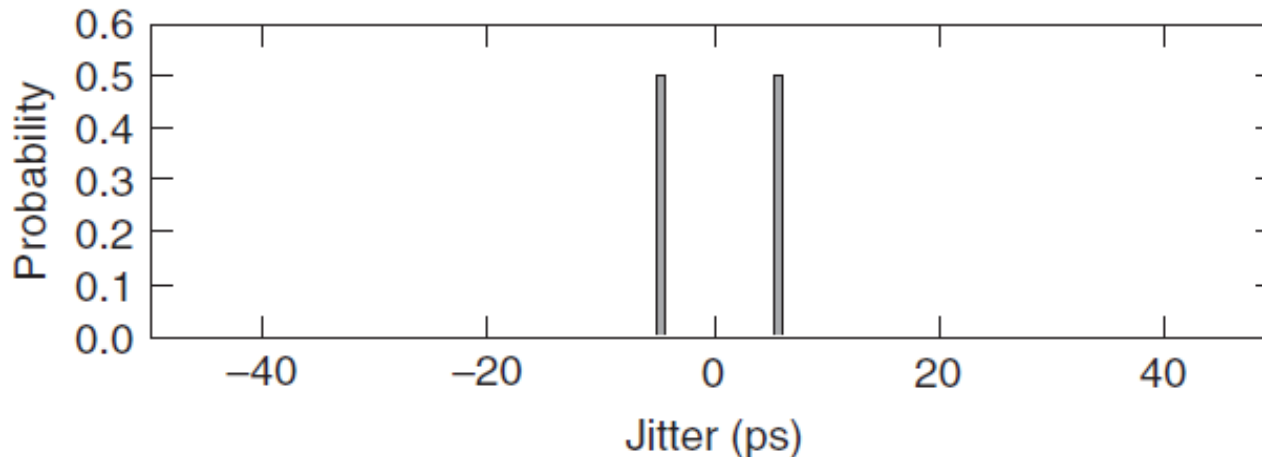
- Data dependent jitter is correlated with either the transmitted data pattern or aggressor (crosstalk) data patterns
- Caused by phenomena such as phase errors in serialization clocks, channel filtering, and crosstalk
- Categories
  - Duty Cycle Distortion (DCD)
  - Intersymbol Interference (ISI)
  - Bounded Uncorrelated Jitter (BUJ)

# Duty Cycle Distortion (DCD)

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- Caused by duty cycle errors in TX serialization clocks and rise/fall delay mismatches in post-serialization buffers
- Resultant PDF from a peak-to-peak duty cycle distortion ( $\alpha_{DCD}$ ) is the sum of two delta functions

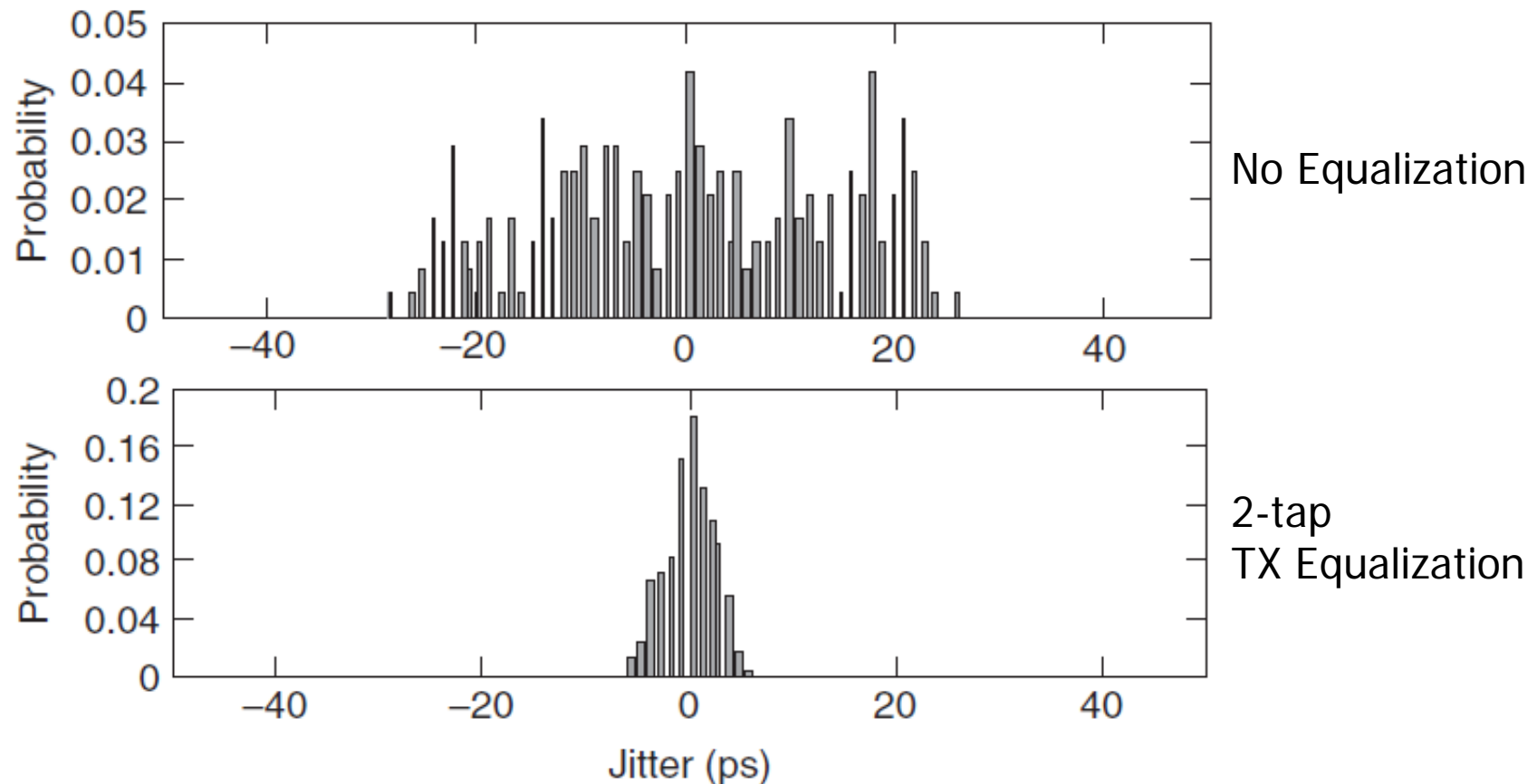
$$PDF_{DCD}(t) = \frac{1}{2} \left[ \delta\left(t - \frac{\alpha_{DCD}}{2}\right) + \delta\left(t + \frac{\alpha_{DCD}}{2}\right) \right]$$





# Intersymbol Interference (ISI)

- Caused by channel loss, dispersion, and reflections
- Equalization can improve ISI jitter



# Bounded Uncorrelated Jitter (BUJ)

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- Not aligned in time with the data stream
- Most common source is crosstalk
- Classified as uncorrelated due to being correlated to the aggressor signals and not the victim signal or data stream
- While uncorrelated, still a bounded source with a quantifiable peak-to-peak value

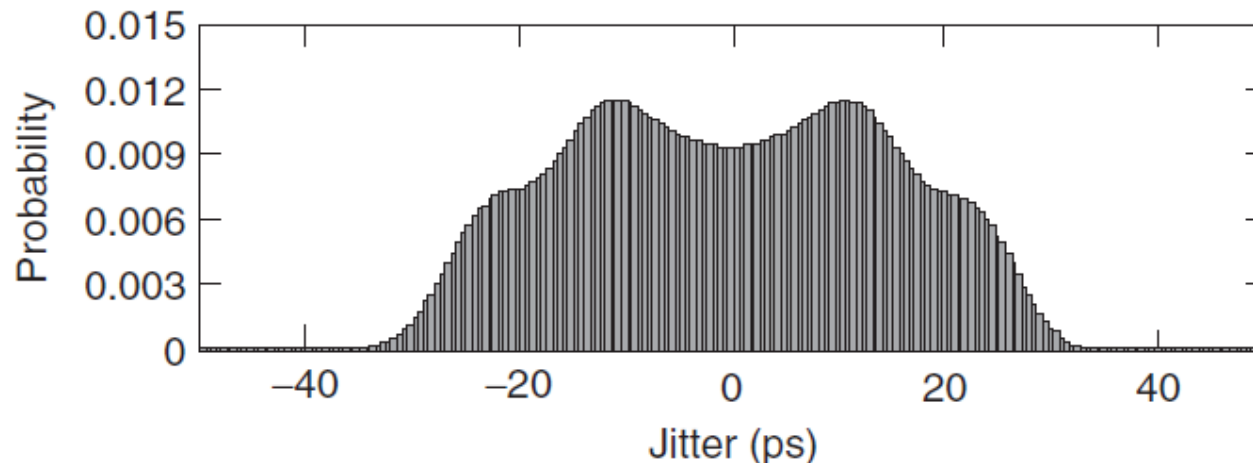
# Total Jitter (TJ)

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- The total jitter PDF is produced by convolving the random and deterministic jitter PDFs

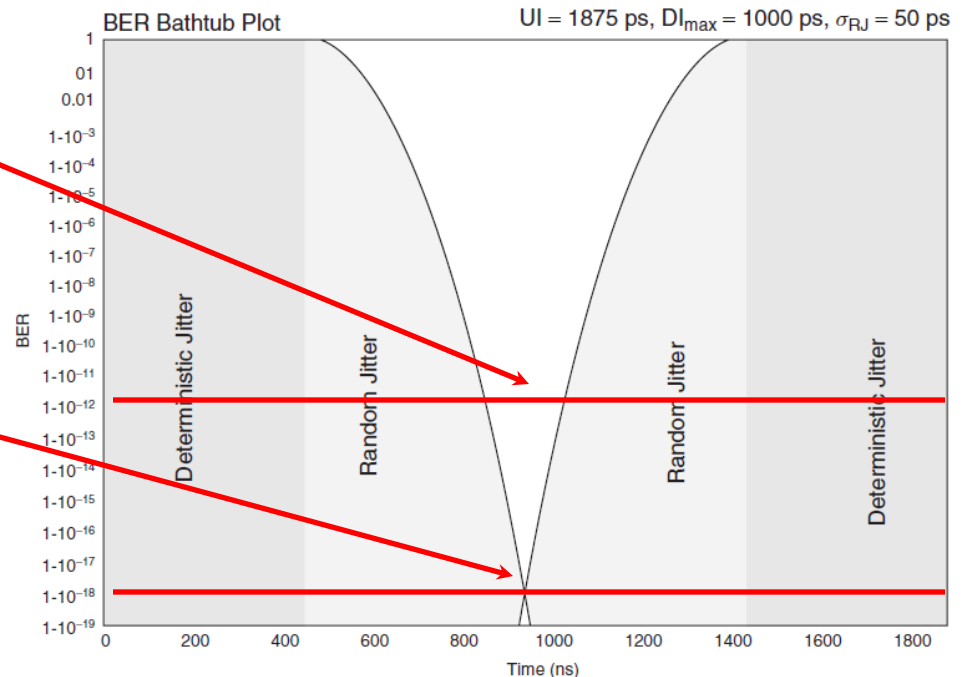
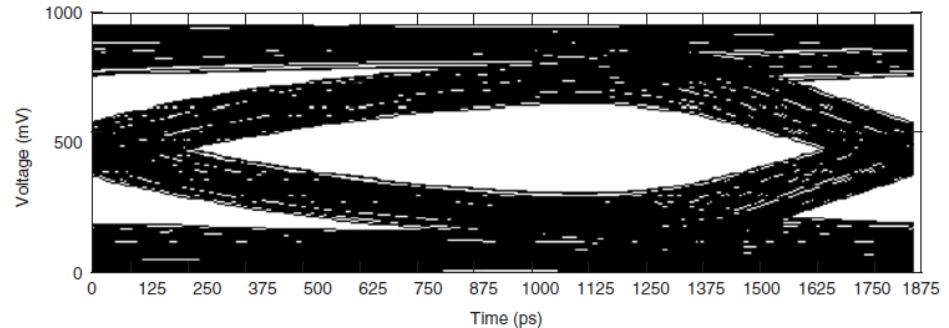
$$PDF_{JT}(t) = PDF_{RJ}(t) * PDF_{DJ}(t)$$

**where**  $PDF_{DJ}(t) = PDF_{SJ}(t) * PDF_{DCD}(t) * PDF_{ISI}(t) * PDF_{BUJ}(t)$



# Jitter and Bit Error Rate

- Jitter consists of both deterministic and **random** components
- Total jitter must be quoted at a given BER
  - At  $BER=10^{-12}$ , jitter  $\sim 1675ps$  and eye width margin  $\sim 200ps$
  - System can potentially achieve  $BER=10^{-18}$  before being jitter limited



# Dual Dirac Jitter Model

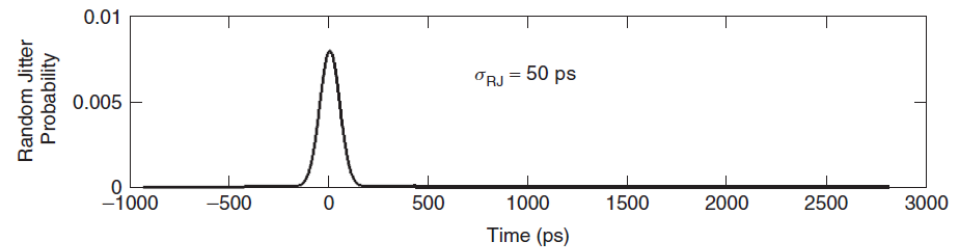
- For system-level jitter budgets, the dual Dirac model allows for the budgeting of deterministic and random jitter components

$$RJ(t) = \frac{1}{\sqrt{2\pi}\sigma_{RJ}} e^{-\frac{t^2}{2\sigma_{RJ}^2}}$$

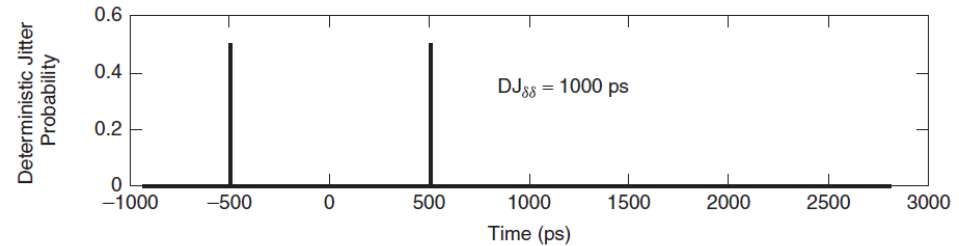
$$DJ(t) = \frac{\delta(t - DJ_{\delta\delta}/2)}{2} + \frac{\delta(t + DJ_{\delta\delta}/2)}{2}$$

$$JT(t) = RJ(t) * DJ(t) = \frac{1}{2\sqrt{2\pi}\sigma_{RJ}} \left[ e^{-\frac{t - DJ_{\delta\delta}/2}{2\sigma_{RJ}^2}} + e^{-\frac{t + DJ_{\delta\delta}/2}{2\sigma_{RJ}^2}} \right]$$

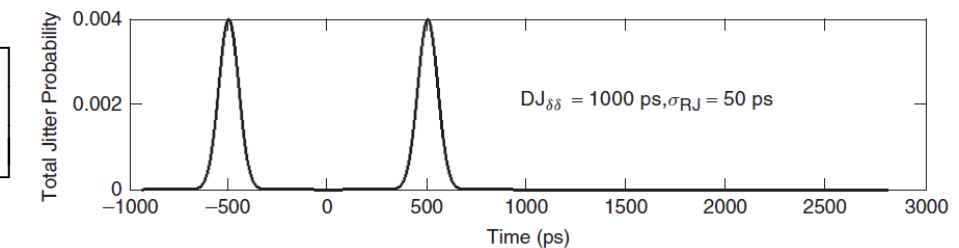
Random Jitter PDF



Deterministic Jitter (dual Dirac) PDF

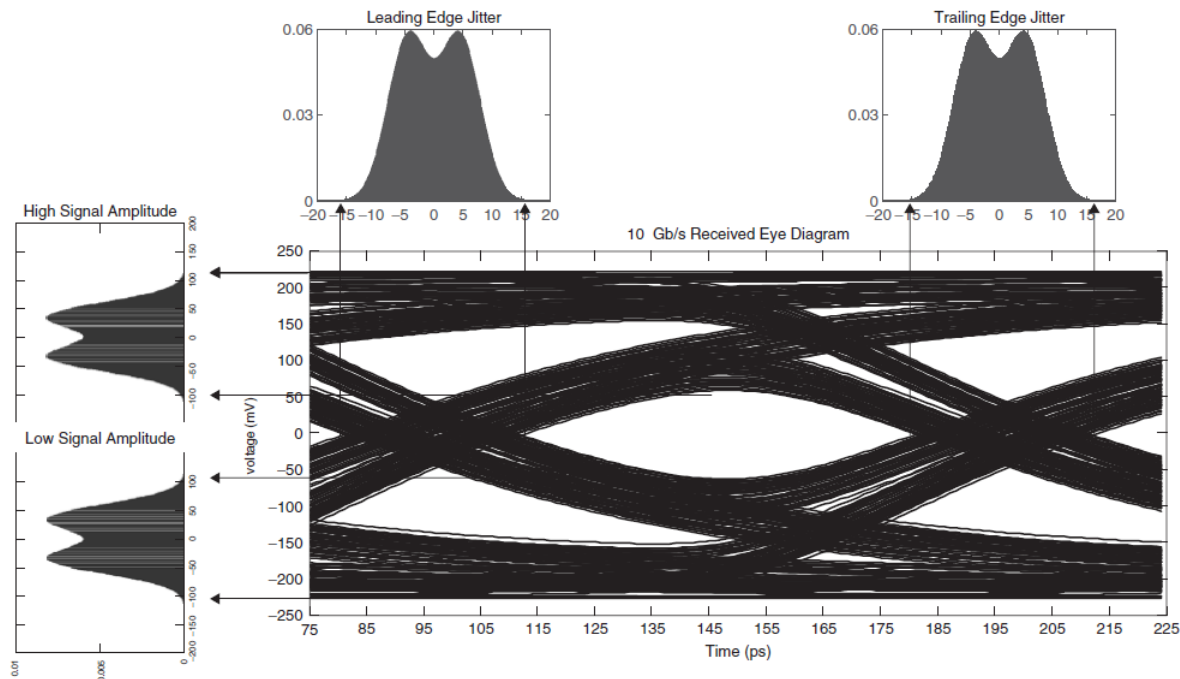


Total Jitter PDF



# Dual Dirac Jitter Model

- Jitter at a given BER is computed considering both leading and trailing edges



$$BER_{lead}(t) = 0.5 \left[ \operatorname{erfc} \left( \frac{t - DJ_{\delta\delta} / 2}{\sqrt{2}\sigma_{RJ}} \right) + \operatorname{erfc} \left( \frac{t + DJ_{\delta\delta} / 2}{\sqrt{2}\sigma_{RJ}} \right) \right], \quad BER_{trail}(t) = 0.5 \left[ \operatorname{erfc} \left( \frac{UI - t - DJ_{\delta\delta} / 2}{\sqrt{2}\sigma_{RJ}} \right) + \operatorname{erfc} \left( \frac{UI - t + DJ_{\delta\delta} / 2}{\sqrt{2}\sigma_{RJ}} \right) \right]$$

$$\text{where } \operatorname{erfc}(t) = \frac{2}{\sqrt{\pi}} \int_t^{\infty} e^{-x^2} dx$$

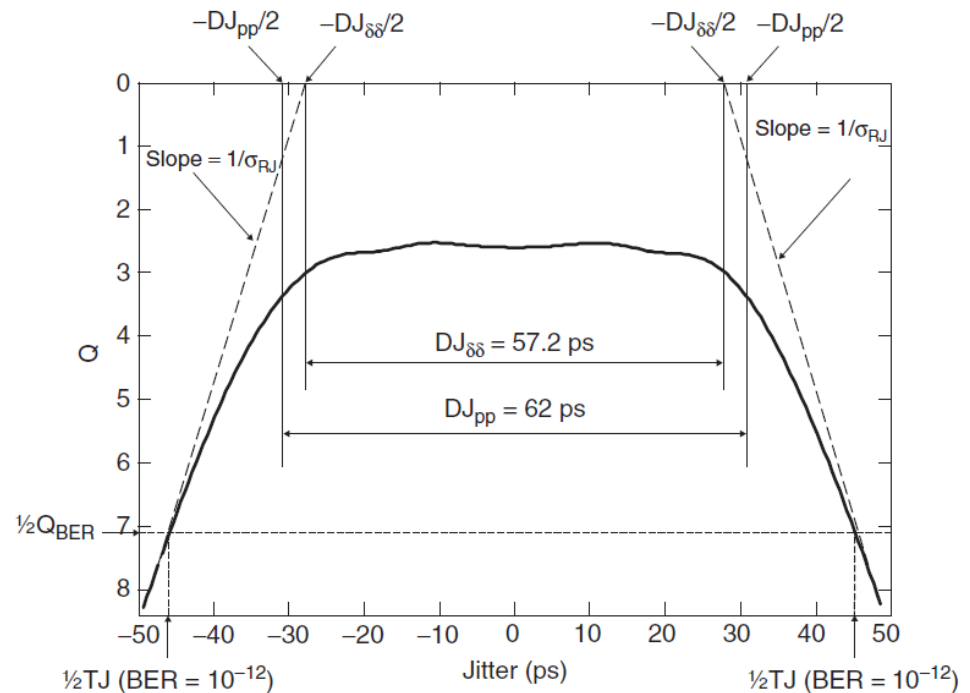
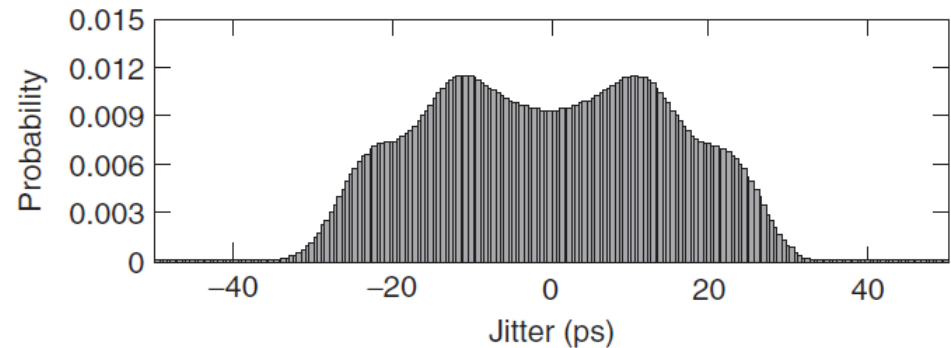
# Dual Dirac Jitter Model Example

- Plot measured jitter PDF vs Q-scale

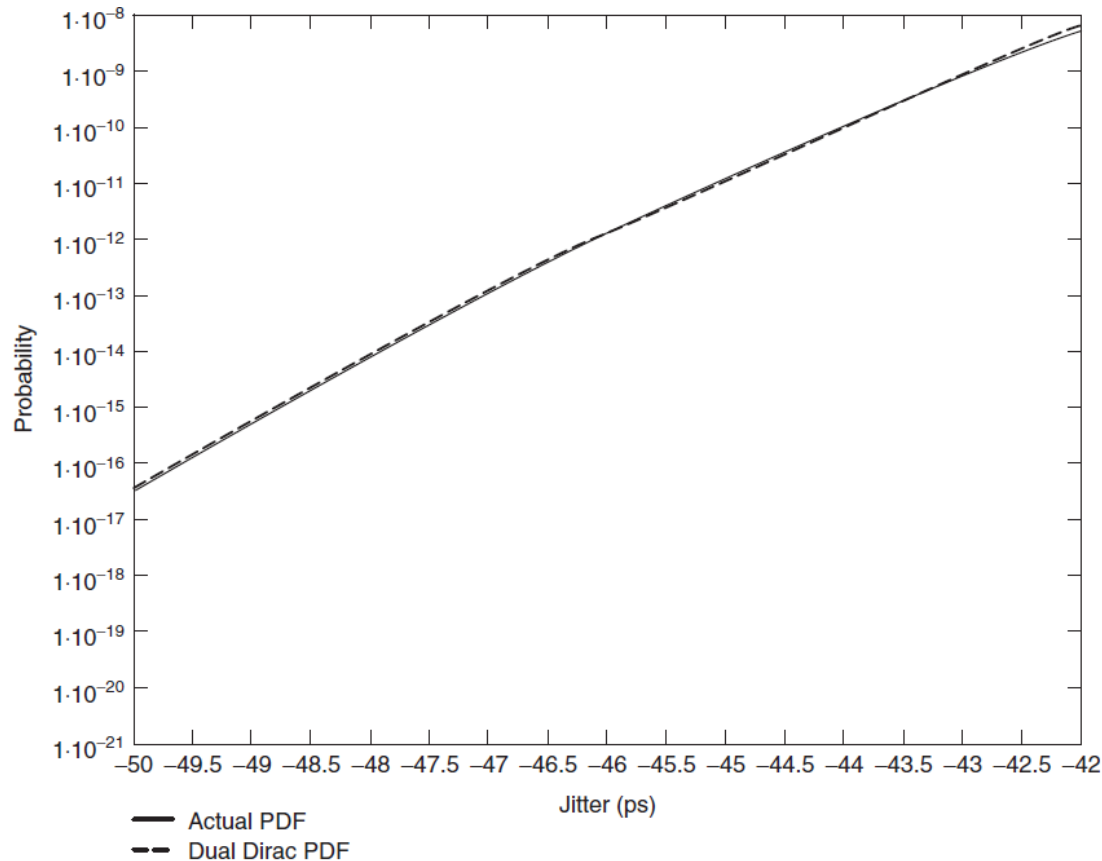
$$Q_{BER}(BER) = \sqrt{2} \operatorname{erf}^{-1} \left( 1 - \frac{BER}{\rho_T} \right)$$

where  $\rho_T$  is the transition density, typically 0.5

- Tails are used to extract  $\sigma_{RJ}$
- Extrapolate to  $Q(0)$  to extract DJ bounds



# Dual Dirac Jitter Model Example



- Extracted dual Dirac model matches well with measured jitter PDF



# Next Time

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- Statistical BER Analysis Tool Overview
- Timing Circuits