

Texas A&M University
Department of Electrical and Computer Engineering

ECEN 325 – Electronics

Fall 2025

Exam #2

Instructor: Sam Palermo

- Please write your name in the space provided below
- Please verify that there are 6 pages in your exam
- Good Luck!

Problem	Score	Max Score
1		10
2		20
3		35
4		35
Total		100

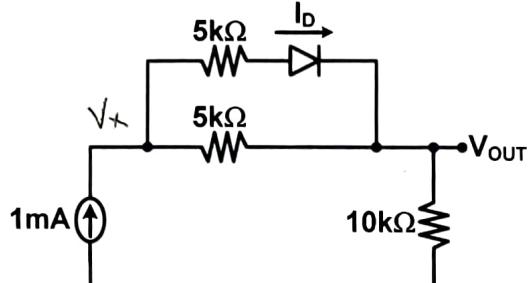
Name: SAM PALERMO

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Problem 1 (10 points)

For all the circuits below, use the constant-voltage-drop diode model ($V_D=0.7\text{V}$), $V_{th}=25.9\text{mV}$, and $n=1$.

a) Find V_{OUT} , I_D , and the small signal diode resistance, r_d .



If Diode is off $\Rightarrow V_{out} = 10V, V_x = 15V$

$V_D = 5V \Rightarrow$ Not consistent

Diode must be forward-biased w/ $V_b = 0.7V$

$$KCL @ V_x$$

$$-1mA + \frac{V_x - 10}{5k} + \frac{V_x - 10.7}{5k} = 0$$

$$2V_x = 25.7V$$

$$V_x = 12.85V$$

$$V_{\text{OUT}} = 10V$$

$$I_D = 430 \mu A$$

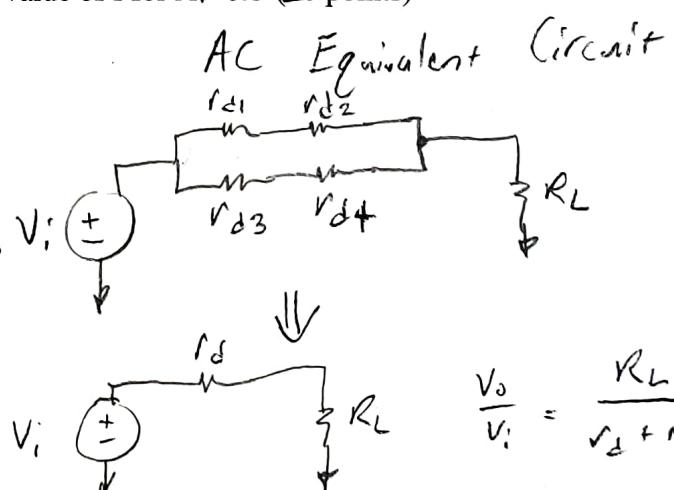
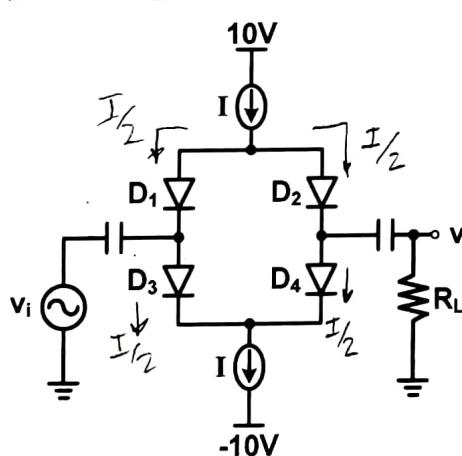
$$r_d = 60.2 \mu$$

Problem 2 (20 points)

Assume for the following circuit that **all diodes are forward biased with equal DC current** and the capacitors act as AC shorts. $V_{th} = 25.9mV$, $\alpha = 1$

Q1) Derive an expression for the small-signal gain $A_v = v_o/v_i$. (10 points)

Q) Assuming $R_L=100\Omega$, calculate the value of I for $A_v=0.8$ (10 points)



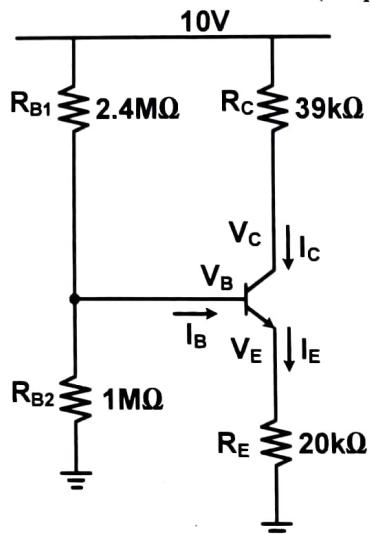
$$\frac{V_o}{V_i} = \frac{R_L}{R_L + R_2}$$

$$r_d = \frac{V+I}{I/2}$$

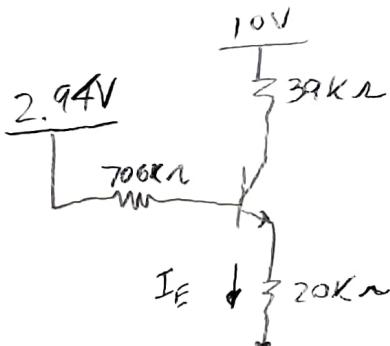
$$\frac{R_L}{\frac{2V_m}{I} + R_L} = 0.8$$

$$I = \frac{2V_{th}}{R_L - R_o} = \frac{2(25.9mV)}{102 - 100} = 2.02mA$$

Problem 3 (35 points)

Assume for Problem 2 that the transistor $\beta=150$, $V_{BE}=0.7V$, and $V_{th}=25.9mV$.a) Calculate the DC values for V_C , V_B , V_E , I_C , I_B , and I_E . Compute the AC small signal parameters g_m , r_π , r_e . (25 points)b) Give the maximum value of R_C such that the transistor remains in the active mode with a minimum $V_{CE}=300mV$. (10 points)

Thevenin Eq Circuit



$$I_E = \frac{2.94V - 0.7V}{20k + \frac{706k}{151}} = 90.8\mu A$$

$$I_C = 90.2\mu A$$

$$I_B = 0.6\mu A$$

$$I_E = 90.8\mu A$$

$$V_C = 6.48V$$

$$V_B = 2.52V$$

$$V_E = 1.82V$$

$$g_m = 348mA/V$$

$$r_\pi = 43.2k\Omega$$

$$r_e = 285\Omega$$

$$I_B = \frac{I_E}{\beta+1} = \frac{90.8\mu A}{151} = 0.6\mu A$$

$$\text{Max } R_C (\text{Active Mode}) = 87.4k\Omega$$

$$I_C = \beta I_B = 150(0.6\mu A) = 90.2\mu A$$

$$V_E = I_E R_E = 90.8\mu A (20k) = 1.82V$$

$$V_B = V_E + 0.7V = 2.52V$$

$$V_C = V_{CC} - I_C R_C = 10V - 90.2\mu A (39k) = 6.48V$$

$$g_m = \frac{I_C}{V_{th}} = \frac{90.2\mu A}{25.9mV} = 348mA/V$$

$$r_\pi = \frac{V_{th}}{I_B} = \frac{25.9mV}{0.6\mu A} = 43.2k\Omega$$

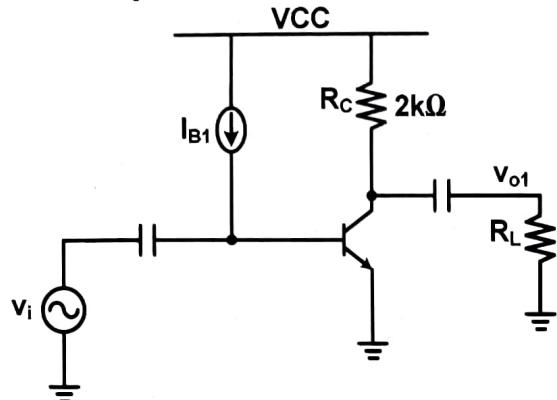
$$r_e = \frac{V_{th}}{I_E} = \frac{25.9mV}{90.8\mu A} = 285\Omega$$

$$\text{Max } R_C: \text{ Min } V_C = V_E + 0.3V = 2.12V$$

$$V_C = V_{CC} - I_C R_C \Rightarrow R_C = \frac{V_{CC} - V_C}{I_C} = \frac{10V - 2.12V}{90.2\mu A} = 87.4k\Omega$$

Problem 4 (35 points)

a) For the following amplifier, assume the transistor $\beta=150$, $r_o=\infty$, and that the transistor is biased such that $g_m=50\text{mA/V}$, $r_e=19.8\Omega$, and $r_\pi=3\text{k}\Omega$. Calculate the amplifier gain, $A_{v1}=v_{o1}/v_i$, for two load resistor conditions: $R_L=\infty$ and $R_L=50\Omega$. Recall that an ideal current source has infinite output resistance, ie $R_B=\infty$. Assume that the capacitors act as AC shorts. (6 points)



$$A_V = -g_m (R_c \parallel R_L)$$

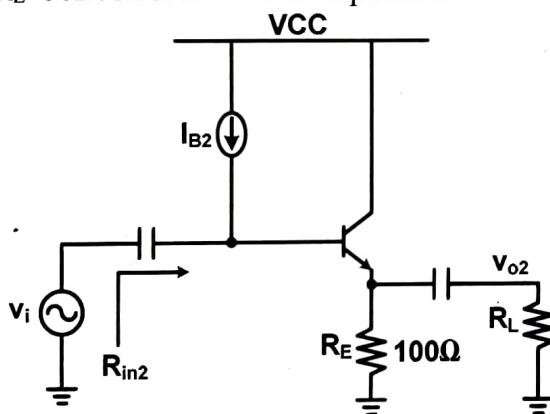
$$w/R_L = \infty; -50\text{mA/V} / (2k) = -100$$

$$w/R_L = 50; -50\text{mA/V} / (2k \parallel 50) = -2.44$$

$$A_{v1} (R_L=\infty) = -100 \text{ V/V}$$

$$A_{v1} (R_L=50\Omega) = -2.44 \text{ V/V}$$

b) For the following amplifier, assume the same transistor bias conditions: $\beta=150$, $r_o=\infty$, and that the transistor is biased such that $g_m=1.16\text{A/V}$, $r_e=0.86\Omega$, and $r_\pi=130\Omega$. Calculate the amplifier gain, $A_{v2}=v_{o2}/v_i$, and the input resistance, R_{in2} , for two load resistor conditions: $R_L=\infty$ and $R_L=50\Omega$. Assume that the capacitors act as AC shorts. (14 points)



$$A_V = \frac{R_L \parallel R_L}{r_e + R_E \parallel R_L}$$

$$w/R_L = \infty \quad A_V = \frac{100}{0.86 + 100} = 0.991$$

$$w/R_L = 50 \quad A_V = \frac{100 \parallel 50}{0.86 + 100 \parallel 50} = 0.975$$

$$A_{v2} (R_L=\infty) = 0.991 \text{ V/V}$$

$$A_{v2} (R_L=50\Omega) = 0.975 \text{ V/V}$$

$$R_{in2} (R_L=\infty) = 15.2 \text{ k}\Omega$$

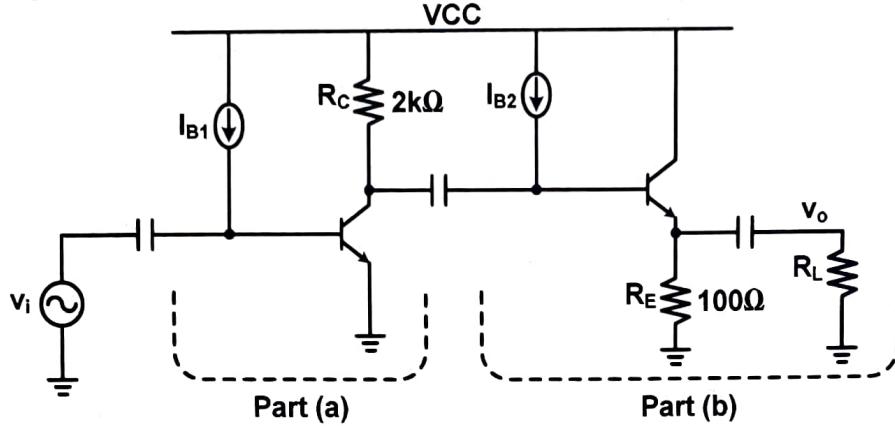
$$R_{in2} (R_L=50\Omega) = 5.16 \text{ k}\Omega$$

$$R_{in2} = r_\pi + (\beta + 1)(R_E \parallel R_L)$$

$$w/R_L = \infty \quad (130 + 151)(100) = 15.2 \text{ k}\Omega$$

$$w/R_L = 50 \quad 130 + 151(100 \parallel 50) = 5.16 \text{ k}\Omega$$

c) The circuit below consists of the amplifier from part (a) cascaded with the amplifier from part (b). Calculate the total 2-stage amplifier gain, $A_v = v_o/v_i$, for two load resistor conditions: $R_L = \infty$ and $R_L = 50\Omega$. Remember to consider the loading of the second stage input resistance on the first stage. Assume that the capacitors act as AC shorts. (15 points)



$$A_V = -g_m (R_C \parallel R_{in2}) \quad \frac{R_E \parallel R_L}{R_C + R_E \parallel R_L}$$

$$w/ R_L = \infty \quad - 50mA/V (2k \parallel 15.2k) (0.991) = -87.6$$

$$w/ R_L = 50 \quad - 50mA/V (2k \parallel 5.16k) (0.915) = -70.3$$

$$A_v (R_L = \infty) = -87.6 \text{ V/V}$$

$$A_v (R_L = 50\Omega) = -70.3 \text{ V/V}$$