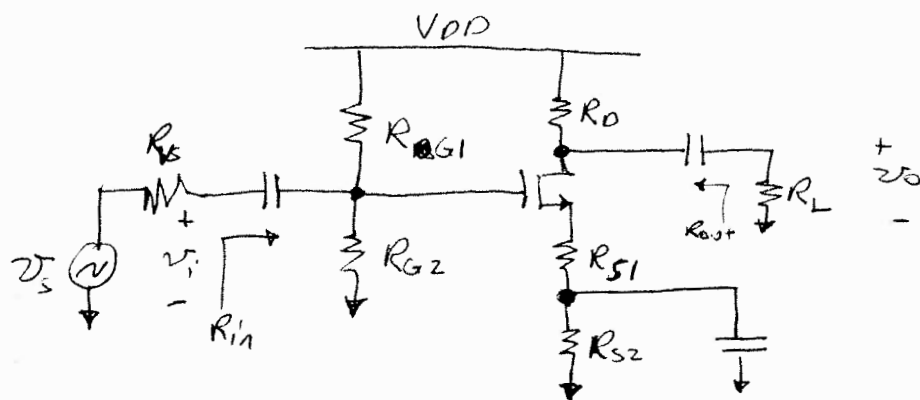


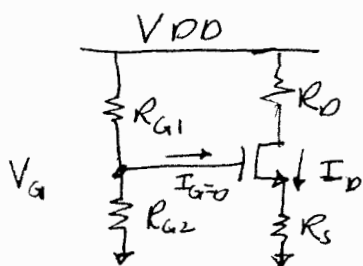
## Common Source w/ Source Resistance



\* Why  $R_S = R_{S1} + R_{S2}$ ?

The source resistor provides a DC bias point more robust to variations ( $K_P$ ,  $V_T$ , temperature). This comes at the cost of reduced gain.

\* DC Biasing



$$V_G = V_{DD} \left( \frac{R_{G2}}{R_{G1} + R_{G2}} \right)$$

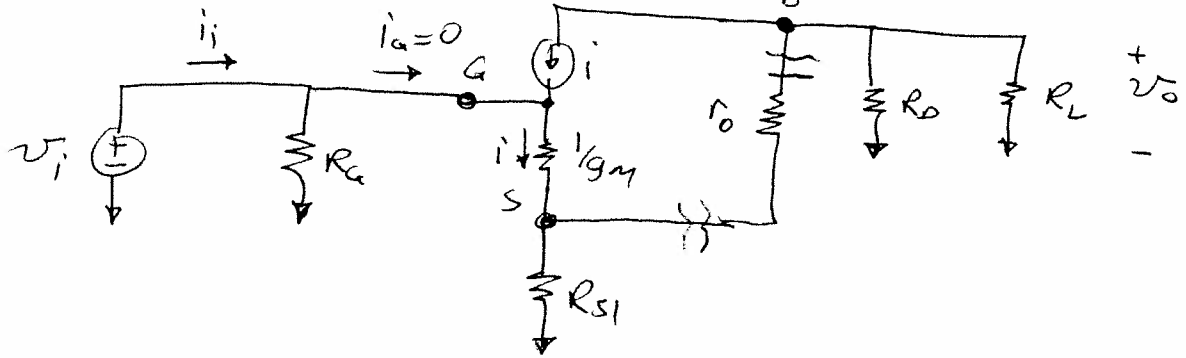
$$I_D = \frac{1}{2} K_P \frac{W}{L} \left[ V_G - I_D R_S - V_{TN} \right]^2$$

⇒ Must solve quadratic equation for  $I_D$   
After some algebra....

$$I_D^2 R_S^2 - I_D \left[ 2(V_G - V_{TN}) R_S + \frac{2}{K_P \frac{W}{L}} \right] + (V_G - V_{TN})^2 = 0$$

⇒ Get 2 solutions, choose the one consistent w/ saturation  
(smaller value)

# AC Equivalent Model (w/T model)



Neglecting  $r_o$  to simplify analysis

$$v_o = -i (R_D \parallel R_L)$$

$$i = \frac{v_i}{\frac{1}{g_m} + R_S}$$

$$v_o = - \frac{v_i (R_D \parallel R_L)}{\frac{1}{g_m} + R_S} = - \frac{g_m (R_D \parallel R_L)}{1 + g_m R_S} v_i$$

$$A_v = \frac{v_o}{v_i} = - \frac{g_m (R_D \parallel R_L)}{1 + g_m R_S}$$

$A_v$  reduced by  $\frac{1}{1 + g_m R_S}$

If  $g_m R_S \gg 1 \Rightarrow$  gain approaches  $-\frac{R_D}{R_S}$   
(and  $R_L$  big)

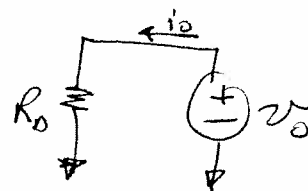
Source Cap allows for both stable DC bias and high AC gain.

\* For  $R_{in}$ , since  $i_G = 0$

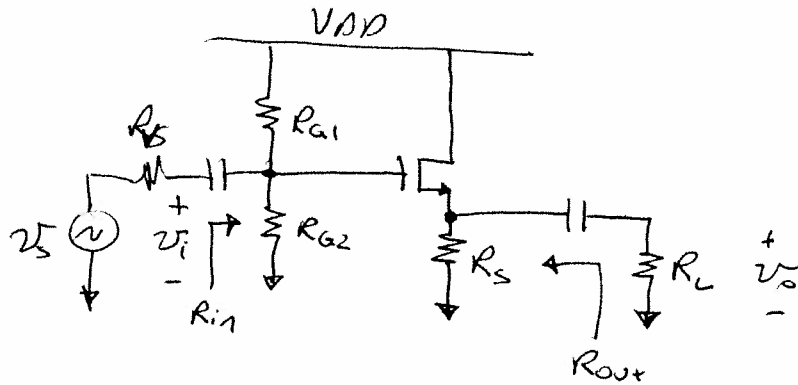
$$R_{in} = \frac{v_i}{i_i} = R_G \quad (\text{unaffected by } R_S)$$

\* For  $R_{out}$ : w/  $v_i = 0, i = 0$

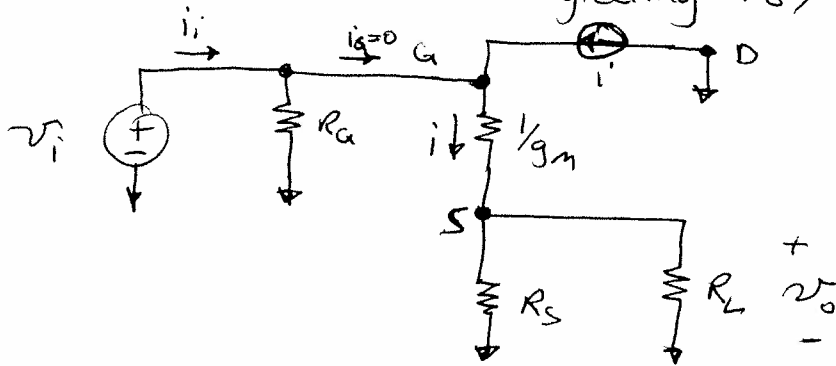
(unaffected by  $R_S$ )  $R_{out} = R_D$



# Common Drain



AC Equivalent Circuit (neglecting  $r_o$ )

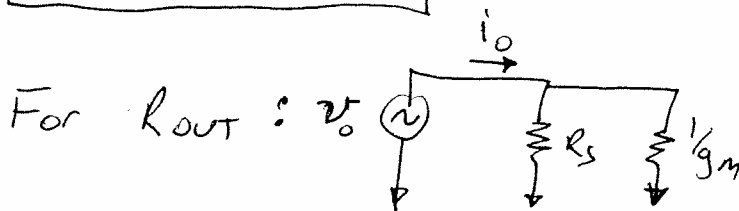


$$v_o = i (R_S \parallel R_L)$$

$$= \frac{v_i (R_S \parallel R_L)}{\frac{1}{g_m} + (R_S \parallel R_L)}$$

$$A_v = \frac{v_o}{v_i} = \frac{g_m (R_S \parallel R_L)}{1 + g_m (R_S \parallel R_L)} \quad (< 1)$$

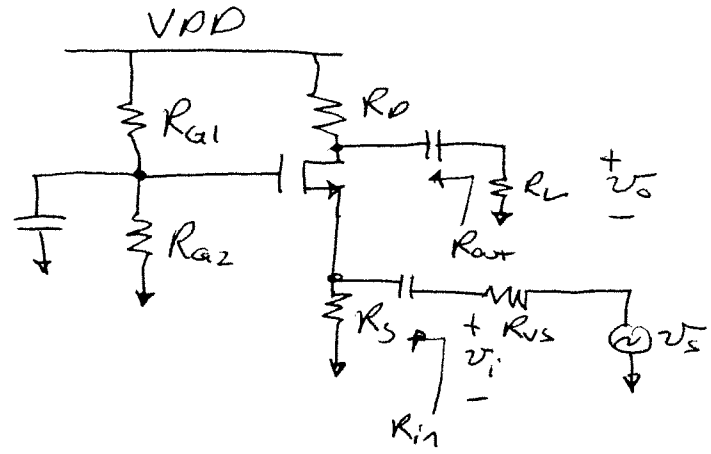
$$R_{in} = \frac{v_i}{i_i} = R_G \quad (\text{same as CS})$$



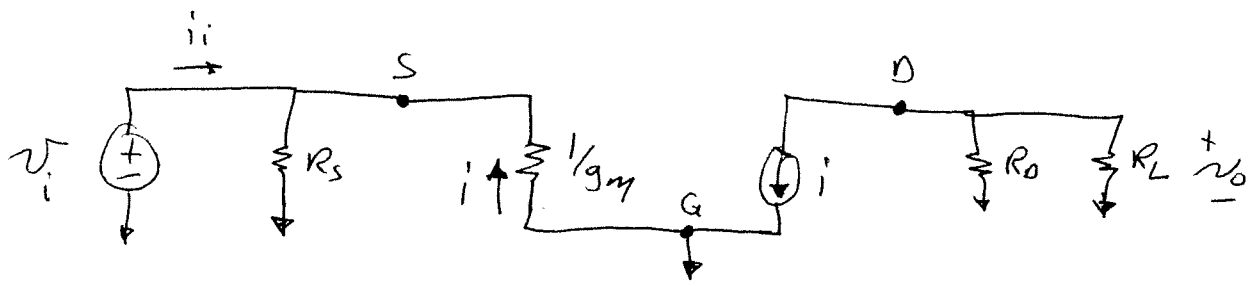
$$R_{out} = R_S \parallel \frac{1}{g_m} \quad (\text{low})$$

$$\approx \frac{1}{g_m}$$

Common Gate



AC Equivalent Circuit (Neglecting  $r_o$ )



$$v_o = -i (R_D \parallel R_L)$$

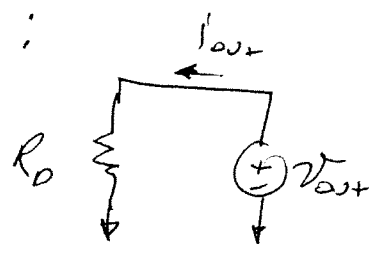
$$i = -\frac{v_i}{1/g_m}$$

$$A_v = \frac{v_o}{v_i} = g_m (R_D \parallel R_L)$$

(Medium Gain)  
(Non-Inv)

$$R_{in} = R_S \parallel \frac{1}{g_m} \approx \frac{1}{g_m} \quad (\text{low})$$

For  $R_{out}$  :



$$R_{out} = R_D \quad (\text{High})$$